



DESCRIPTION

The MP1925 is a high-frequency, half-bridge, N-channel power MOSFET driver. Its low-side and high-side driver channels are controlled independently and matched with less than 5ns of time delay. Under-voltage lockout (UVLO) on both the high-side and low-side supplies forces the outputs low in the event that the supply is insufficient. The integrated bootstrap diode reduces the external component count.

MP1925 is available in a QFN-8 (4mmx4mm) package.

FEATURES

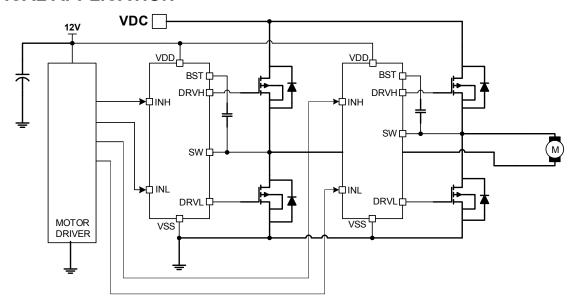
- Drives an N-Channel MOSFET Half-Bridge
- 115V Bootstrap Voltage Range
- On-Chip Bootstrap Diode
- Typical Propagation Delay of 20ns
- Gate Driver Matching of Less than 5ns
- Drives a 2.2nF Load with 15ns of Rise Time and 10ns of Fall Time at 12V VDD
- TTL-Compatible Input
- Quiescent Current of Less than 150µA
- UVLO for Both High-Side and Low-Side **Gate Drivers**
- Available in a QFN-8 (4mmx4mm) Package

APPLICATIONS

- **Motor Drivers**
- Telecom Half-Bridge Power Supplies
- Avionics DC/DC Converters
- Two-Switch Forward Converters
- **Active Clamp Forward Converters**

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TYPICAL APPLICATION



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ORDERING INFORMATION

Part Number*	Package	Top Marking
MP1925HR	QFN-8 (4mmx4mm)	See Below

* For Tape & Reel, add suffix –Z (e.g. MP1925HR–Z) For RoHS compliant packaging, add suffix –LF (e.g. MP1925HR–LF–Z)

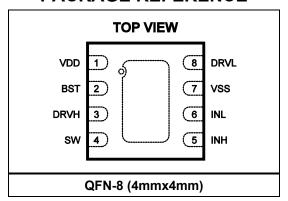
TOP MARKING

MPSYWW MP1925

LLLLLL

MPS: MPS prefix Y: Year code WW: Week code MP1925: Part number LLLLL: Lot number

PACKAGE REFERENCE



PIN FUNCTIONS

Pin#	Name	Description			
1	VDD	Supply input. VDD supplies power to the internal circuitry. Place a decoupling apacitor to ground close to VDD to ensure a stable and clean supply.			
2	BST	Bootstrap. BST is the positive power supply for the internal floating high-side MOSFET driver. Connect a bypass capacitor between BST and SW.			
3	DRVH	Floating driver output.			
4	SW	Switching node.			
5	INH	Control signal input for the floating driver.			
6	INL	Control signal input for the low-side driver.			
7	VSS, exposed pad	Chip ground. Connect the exposed pad to VSS for proper thermal operation.			
8	DRVL	Low-side driver output.			



ABSOLUTE MAXIMUM RATINGS (1)

ABOULUIL IIIAAIIIIOIII IA	A111100
Supply voltage (V _{DD})	-0.3V to +18V
SW voltage (V _{SW})	
SW voltage (V _{SW})25V(<10	
BST voltage (V _{BST})	0.3V to +115V
BST voltage (V _{BST})15V(<10	Ons) to +115V
BST to SW	-0.3V to +18V
DRVH to SW (2)	
DRVH to SW (2)5V(<10	
DRVH to VSS0.3V to (BS1	- SW) + 0.3V
DRVH to VSS	T 1/00) - 0 01/
15V (<100ns) to (BS	S1-VSS)+0.3V
DRVL to VSS (2)	
DRVL to VSS (2)5V(<10	
INH/NL to VSS0.3V t	$0 (V_{DD} + 0.3V)$
INH/INL to VSS	() (DD + 0.0) ()
5V(<100ns) to	(VDD + 0.3V)
All other pins0.3V t	$(0 (V_{DD} + 0.3V)$
Continuous power dissipation (T _A	
QFN-8 (4mmx4mm)	
Junction temperature	
Lead temperature	200 C
Storage temperature6	5 C 10 + 150 C
Recommended Operating Co	nditions (4)
Supply voltage (V _{DD})	8.0V to 15.0V
SW voltage (V _{SW})	1.0V to +100V
SW slew rate	
Operating junction temp $(T_J = T_A)$	
4	0°C to +125°C

Thermal Resistance (5)	$oldsymbol{ heta}_{JA}$	$oldsymbol{ heta}_{JC}$	
QFN-8 (4mmx4mm)	47	7	.°C/W

Notes:

- 1) Exceeding these ratings may damage the device. The repetitive pulse rating is guaranteed for period of 100ns or less with a maximum repetition rate of 1000kHz when VDD is 15V or less.
- DRVH and DRVL are outputs pins, cannot be connected to external supply voltage.
- The maximum allowable power dissipation is a function of the maximum junction temperature T_J (MAX), the junction-to-ambient thermal resistance $\theta_{JA},$ and the ambient temperature T_A. The maximum allowable continuous power dissipation at any ambient temperature is calculated by PD (MAX) = (TJ (MAX) - T_A) / θ_{JA}. Exceeding the maximum allowable power dissipation produces an excessive die temperature, causing the regulator to go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- The device is not guaranteed to function outside of its operating conditions.
- Measured on JESD51-7, 4-layer PCB.



ELECTRICAL CHARACTERISTICS

 V_{DD} = V_{BST} - V_{SW} = 12V, V_{SS} = V_{SW} = 0V, no load at DRVH and DRVL, T_J = -40°C to +125°C, typical value is tested at T_J = 25°C, unless otherwise noted.

Supply Currents IDDQ INL = INH = 0	Parameter	Symbol	Condition	Min	Тур	Max	Units
VDD operating current IDDO fsw = 500kHz 9							
Floating driver quiescent current IBSTQ INL = INH = 0	VDD quiescent current	I _{DDQ}	INL = INH = 0		110	150	μA
Floating driver operating current IBSTO ISW = 500kHz BST = SW = 100V 0.05 1 μA	VDD operating current	I _{DDO}	f _{SW} = 500kHz		9		mA
Leakage current ILIX BST = SW = 100V 0.05	Floating driver quiescent current	I _{BSTQ}	INL = INH = 0		60	90	μA
Inputs Inputs INL/INH high	Floating driver operating current	I _{BSTO}	f _{SW} = 500kHz		8		mA
INL/INH high	Leakage current	I _{LK}	BST = SW = 100V		0.05	1	μA
INL/INH low	Inputs						
NLI/INH internal pull-down resistance	INL/INH high				2	2.4	V
Tesistance	INL/INH low			1	1.4		V
Page Protection Page Protection Page Protection Page Pa		D			195		kΟ
VDD rising threshold VDDR		ININ			100		N12
VDD hysteresis							
SST-SVV rising threshold VBSTR S.8 6.5 6.9 V	VDD rising threshold	V_{DDR}		6	6.8	7.2	
BST-SW hysteresis VBSTH	VDD hysteresis	V_{DDH}			0.4		V
Bootstrap Diode	BST-SW rising threshold	V _{BSTR}		5.8	6.5	6.9	V
Bootstrap diode VF at 100μA VF1 0.5 V Bootstrap diode VF at 100mA VF2 0.95 V Bootstrap diode dynamic R Rp At 100mA 2.5 Ω Low-Side Gate Driver Useron Side Gate Driver 0.1 V Low-level output voltage Voll Io = 100mA 0.1 V High-level output voltage to rail Vohl Io = -100mA 0.19 V Source current (6) Iolh VDRVL = 0V, VDD = 12V 3 A VDRVL = VDD = 16V 4.7 A VDRVL = VDD = 16V 6 A Floating Gate Driver VOHL Io = 100mA 0.1 V Low-level output voltage VOLH Io = 100mA 0.1 V High-level output voltage to rail VOHH Io = -100mA 0.19 V Source current (6) IoHH VDRVH = 0V, VDD = 12V 2.6 A VDRVH = 0V, VDD = 16V 4 A A	BST-SW hysteresis	V _{BSTH}			0.4		V
Bootstrap diode VF at 100mA VF2	Bootstrap Diode						
Bootstrap diode dynamic R R _D At 100mA 2.5 Ω	Bootstrap diode VF at 100µA	V_{F1}			0.5		V
	Bootstrap diode VF at 100mA	V _{F2}			0.95		V
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	Bootstrap diode dynamic R	R_D	At 100mA		2.5		Ω
High-level output voltage to rail V_{OHL} I_{O} = -100mA 0.19 V_{ORVL} = 0V, V_{DD} = 12V 0.19 V_{DRVL} = 0V, V_{DD} = 12V 0.19 0	Low-Side Gate Driver						
Source current (6) I_{OHL} $V_{DRVL} = 0V, V_{DD} = 12V$ 3 A $V_{DRVL} = 0V, V_{DD} = 16V$ 4.7 A A $V_{DRVL} = 0V, V_{DD} = 16V$ 4.5 A	Low-level output voltage	Voll	I _O = 100mA		0.1		V
Source current (6) $V_{DRVL} = 0V, V_{DD} = 16V$ 4.7 A $V_{DRVL} = 0V_{DD} = 12V$ 4.5 A $V_{DRVL} = V_{DD} = 12V$ 6 A $V_{DRVL} = V_{DD} = 16V$ 6 A $V_{DRVL} = V_{DD} = 16V$ 7 A $V_{DRVL} = V_{DD} = 16V$ 9 A $V_{DRVH} = 0V, V_{DD} = 12V$ 9 A $V_{DRVH} = 0V, V_{DD} = 16V$ 9 A $V_{DRVH} = 0V, V_{DD} = 16V$ 9 A $V_{DRVH} = 0V, V_{DD} = 16V$ 9 A $V_{DRVH} = 0V, V_{DD} = 12V$ 9 A $V_{DRVH} = 0V_{DD} = 12V$ 9 A	High-level output voltage to rail	Vohl	I _O = -100mA		0.19		V
Sink current (6) $I_{OLL} = V_{DRVL} = 0V, V_{DD} = 16V $ $V_{DRVL} = V_{DD} = 12V $ $V_{DRVL} = V_{DD} = 16V $ $V_{DRVL} = V_{DD} = 12V $	Course ourrent (6)	Іонь	$V_{DRVL} = 0V, V_{DD} = 12V$		3		Α
Sink current (6) I_{OLL} $V_{DRVL} = V_{DD} = 16V$ G A A Floating Gate Driver $I_{OWL} = I_{OWL} $	Source current (9)		$V_{DRVL} = 0V, V_{DD} = 16V$		4.7		Α
Floating Gate Driver Low-level output voltage V_{OLH} I_{O} = 100mA V_{OLH} I_{O} = 12V V_{OLH} I_{O} = 12V I_{OLH} I_{OHH} I_{OH	Cial accept (6)		$V_{DRVL} = V_{DD} = 12V$		4.5		Α
Low-level output voltage V_{OLH} $I_O = 100 \text{mA}$ 0.1 V High-level output voltage to rail V_{OHH} $I_O = -100 \text{mA}$ 0.19 V Source current (6) I_{OHH} $V_{DRVH} = 0V, V_{DD} = 12V$ 2.6 A Variety output voltage to rail V_{OHH} $V_{DRVH} = 0V, V_{DD} = 16V$ 4 A Sink current (6) $V_{DRVH} = V_{DD} = 12V$ 4.5 A	Sink current (9)	IOLL	V _{DRVL} = V _{DD} = 16V		6		Α
High-level output voltage to rail V_{OHH} I_{O} = -100mA 0.19 V_{OHH} Source current (6) I_{OHH} V_{DRVH} = 0V, V_{DD} = 12V 2.6 A V_{DRVH} = 0V, V_{DD} = 16V 4 A V_{DRVH} = 0V, V_{DD} = 12V 4.5 A	Floating Gate Driver						
Source current (6) I_{OHH} $V_{DRVH} = 0V, V_{DD} = 12V$ 2.6 A $V_{DRVH} = 0V, V_{DD} = 16V$ 4 A $V_{DRVH} = 0V, V_{DD} = 12V$ 4.5 A	Low-level output voltage	V _{OLH}	I _O = 100mA		0.1		V
Source current (6) $V_{DRVH} = 0V, V_{DD} = 16V$ 4 A Sink current (6) $V_{DRVH} = V_{DD} = 12V$ 4.5 A	High-level output voltage to rail		I _O = -100mA		0.19		V
$V_{DRVH} = 0V, V_{DD} = 16V$ 4 A Sink current (6) $V_{DRVH} = V_{DD} = 12V$ 4.5 A	Course current (6)		$V_{DRVH} = 0V, V_{DD} = 12V$		2.6		Α
SINK CURRENT (0)	Source current (*)	Іонн	V _{DRVH} = 0V, V _{DD} = 16V		4		Α
Sink current $\stackrel{\text{OLH}}{}$ $V_{DRVH} = V_{DD} = 16V$ 5.9 A	Ciple ourrent (6)	1	$V_{DRVH} = V_{DD} = 12V$		4.5		Α
	Sink current (9)	IOLH	$V_{DRVH} = V_{DD} = 16V$		5.9		Α



ELECTRICAL CHARACTERISTICS (continued)

 V_{DD} = V_{BST} - V_{SW} = 12V, V_{SS} = V_{SW} = 0V, no load at DRVH and DRVL, T_J = -40°C to +125°C, typical value is tested at T_J = 25°C, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Switching Specification – Low-Side Gate Driver						
Turn-off propagation delay INL falling to DRVL falling	tolff			20		ns
Turn-on propagation delay INL rising to DRVL rising	t _{DLRR}			20		
DRVL rise time		C _L = 2.2nF		15		ns
DRVL fall time		C _L = 2.2nF		10		ns
Switching Specification – Floating	ng Gate D	river				
Turn-off propagation delay INH falling to DRVH falling	t _{DHFF}			20		ns
Turn-on propagation delay INH rising to DRVH rising	t DHRR			20		ns
DRVH rise time		$C_L = 2.2nF$		15		ns
DRVH fall time		$C_L = 2.2 nF$		10		ns
Switching Specification – Match	ing					
Floating driver turn-off to low-side driver turn-on ⁽⁶⁾	t _{MON}			1	5	ns
Low-side driver turn-off to floating driver turn-on ⁽⁶⁾	t _{MOFF}			1	5	ns
Minimum input pulse width that changes the output ⁽⁶⁾	t _{PW}				50	ns
Bootstrap diode turn-on or turn-off time ⁽⁶⁾	t BS			10		ns
Thermal shutdown				150		°C
Thermal shutdown hysteresis				25		°C

Note:

6) Guaranteed by design.

TIMING DIAGRAM

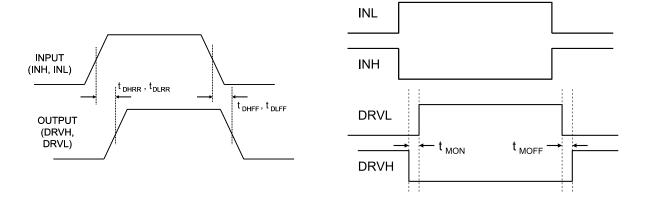
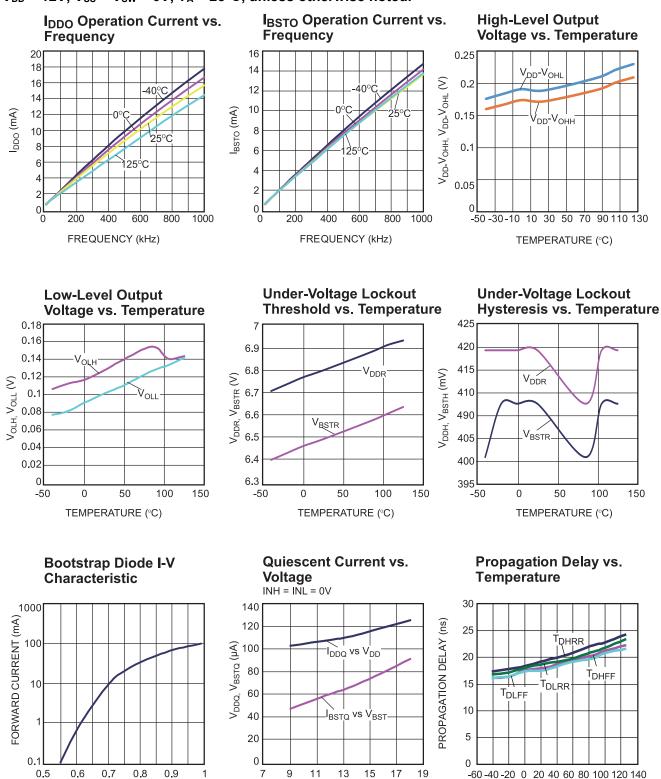


Figure 1: Timing Diagram



TYPICAL CHARACTERISTICS

 V_{DD} = 12V, V_{SS} = V_{SW} = 0V, T_A = 25°C, unless otherwise noted.



FORWARD VOLTAGE (V)

 $V_{DD,}V_{BST}(V)$

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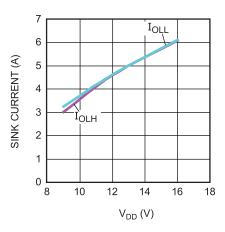
TEMPERATURE (°C)



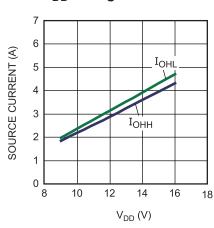
TYPICAL CHARACTERISTICS (continued)

 V_{DD} = 12V, V_{SS} = V_{SW} = 0V, T_A = 25°C, unless otherwise noted.

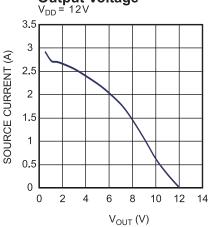
Sink Current vs. V_{DD} Voltage



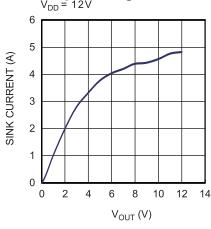
Source Current vs. V_{DD} Voltage



Source Current vs.
Output Voltage
VDD = 12V



Sink Current vs. Output Voltage V_{DD} = 12V



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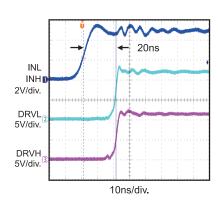
TYPICAL PERFORMANCE CHARACTERISTICS

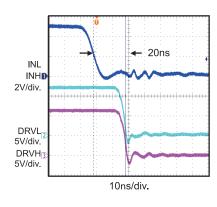
 V_{DD} = 12V, V_{SS} = V_{SW} = 0V, T_A = 25°C, unless otherwise noted.

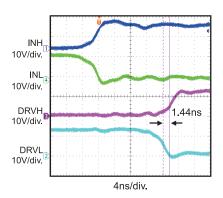
Turn-On Propagation Delay

Turn-Off Propagation Delay

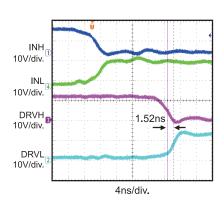
Gate Drive Matching t_{MOFF}



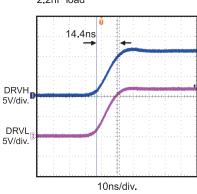




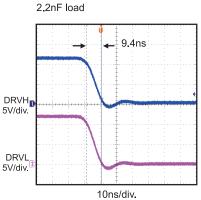
Gate Drive Matching t_{MON}



Drive Rise Time 2.2nF load



Drive Fall Time





FUNCTIONAL BLOCK DIAGRAM

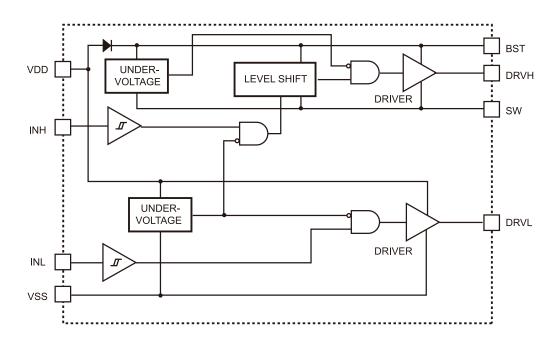


Figure 2: Functional Block Diagram



APPLICATION INFORMATION

The input signals of INH and INL can be controlled independently. If both INH and INL control the high-side and low-side MOSFETs of the same bridge, set a sufficient dead time

between INH and INL low (and vice versa) to avoid shoot-through (see Figure 3). Dead time is defined as the time interval between INH low and INL low.

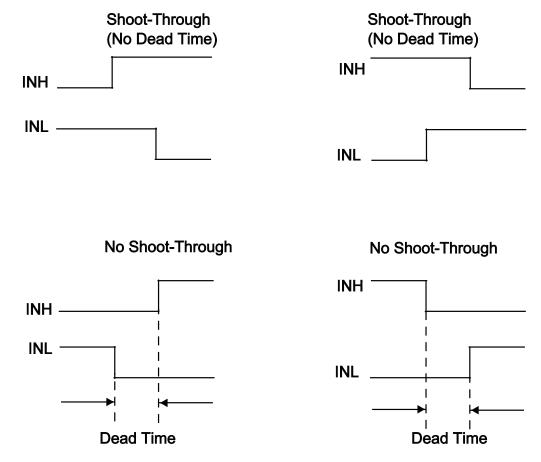


Figure 3: Shoot-Through Timing Diagram



REFERENCE DESIGN CIRCUITS

Half-Bridge Converter

The MP1925 drives the MOSFETS via alternating signals with dead time in half-bridge converter topology. The input voltage can rise up

to 100V with the alternating signals (INT and INL) coming from the PWM controller (see Figure 4).

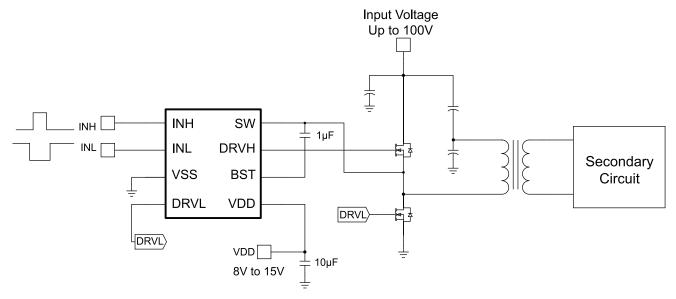


Figure 4: Half-Bridge Converter

Two-Switch Forward Converter

In two-switch forward converter topology, both MOSFETs turn on and off simultaneously. The input signals (INH and INL) come from a PWM controller that senses the output voltage and output current during current mode control.

The Schottky diodes clamp the reverse swing of the power transformer, and must be rated for the input voltage. The input voltage can rise up to 100V (see Figure 5).

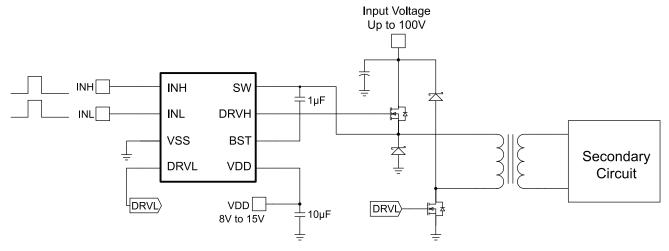


Figure 5: Two-Switch Forward Converter

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Active Clamp Forward Converter

In active clamp forward converter topology, the MP1925 drives the MOSFETs with alternating signals. The high-side MOSFET, in conjunction with C_{reset} , is used to reset the power transformer without loss.

This topology is optimal for running at duty cycles exceeding 50%. The device may not be able to run at 100V in this topology (see Figure 6).

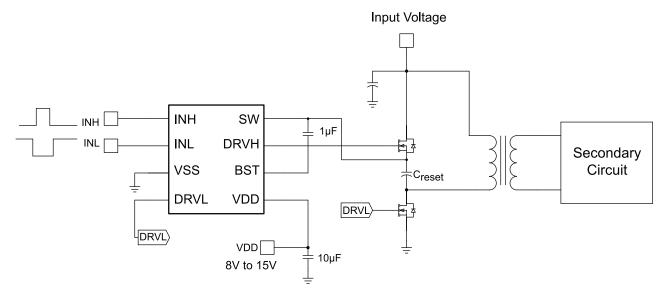
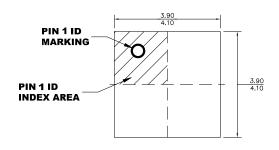


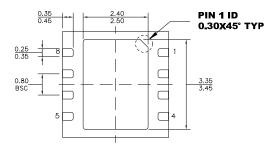
Figure 6: Active Clamp Forward Converter



PACKAGE INFORMATION

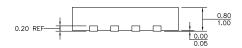
QFN-8 (4mmx4mm)



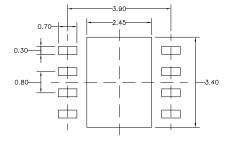


TOP VIEW

BOTTOM VIEW



SIDE VIEW



NOTE:

- 1) ALL DIMENSIONS ARE IN MILLIMETERS. 2) EXPOSED PADDLE SIZE DOES NOT INCLUDE MOLD FLASH.
- 3) LEAD COPLANARITY SHALL BE 0.08 MILLIMETERS MAX.
- 4) JEDEC REFERENCE IS MO-220. 5) DRAWING IS NOT TO SCALE.

RECOMMENDED LAND PATTERN



Revision History

Revision #	Revision Date	Description	Pages Updated
1.01	07/24/2020	Update transient negative Absolute Maximum Ratings	Page 3

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